## Introduction to OFDM Systems





#### Outline

- OFDM Overview
- OFDM System Model
- Orthogonality
- Multi-carrier Equivalent Implementation by Using IDFT (IFFT)
- Cyclic Prefix (CP)





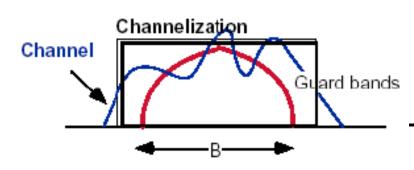
#### OFDM

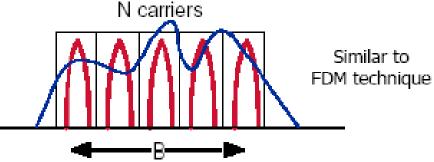
- Orthogonal Frequency Division Multiplexing
- Frequency Division Multiplexing (FDM) or multi-tone systems have been employed in military applications since the 196Os.

 OFDM employs multiple carriers overlapping in the frequency domain.



Single carrier (SC) vs. multi-carrier (MC)





- Single carrier: data are transmitted over only one carrier
- Selective fading

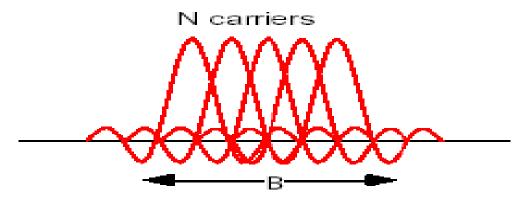
- Multi-carrier: data are shared among several carriers and simultaneously transmitted
- Flat fading per subcarrier



- The basic principle of OFDM is to split a high-rate data stream into a number of lower rate streams that are transmitted simultaneously over a number of sub-carriers.
- \* It eliminates or alleviates the problem of <u>multi-path</u> channel fading effect, <u>low spectrum efficiency</u>, and frequency selective fading.



#### OFDM modulation



Carriers are mutually orthogonal

- Features
  - No intercarrier guard bands
  - Overlapping of bands
  - Spectral efficiency
  - Easy implementation by IFFTs
  - Very sensitive to synchronization



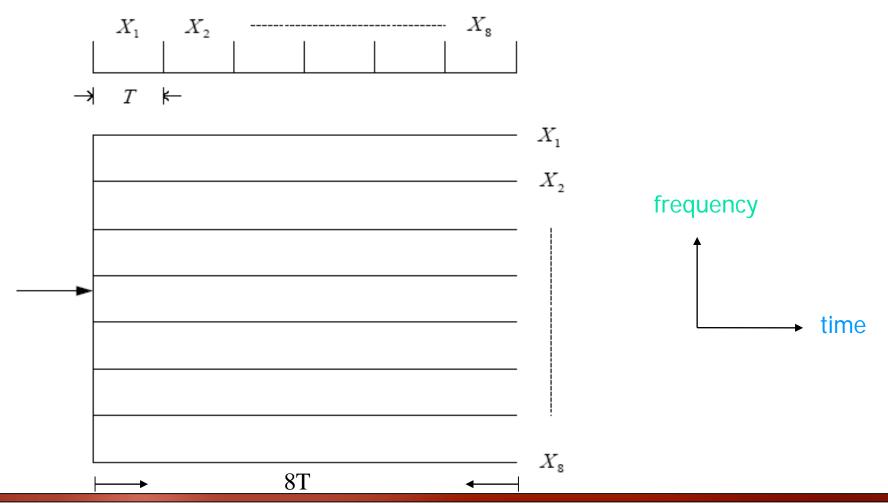
# Applications of OFDM Technology

- Broadband Wired Access: Asymmetric
   Digital Subscriber Loop (ADSL), Digital Multi-tone (DMT).
- Wireless LANs (IEEE 802.11a/g, IEEE 802.11n, HIPERLAN-2)
- Digital Broadcasting (DAB, DVB-T, DVB-H)
- WiMAX (IEEE 802.16 Series), 3GPP Long Term Evolution (3GPP LTE), 4G.
- Wireless Personal Area Network (WPAN): IEEE 802.15a/MBOA
- Power Line





#### Multi-carrier Block Transmission

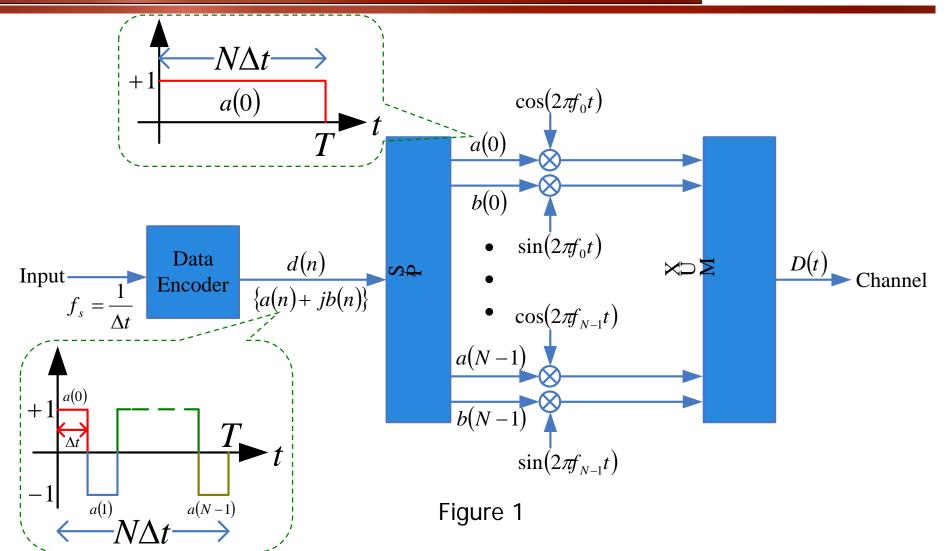




- \* OFDM: A block modulation scheme that transmits a block N source symbols in parallel by using subcarriers
  - \* Sub-carriers are orthogonal in time, but overlapped in frequency.
  - Frequency spacing:  $\Delta f = \frac{1}{T_{FFT}}$

$$\int_0^{T_{FFT}} \cos(2\pi f_1 t) \cos(2\pi (f_1 + \Delta f)t) dt = 0$$







- \* An OFDM system transmitter shown in Figure 1.
- \* The transmitted waveform D(t) can be expressed as

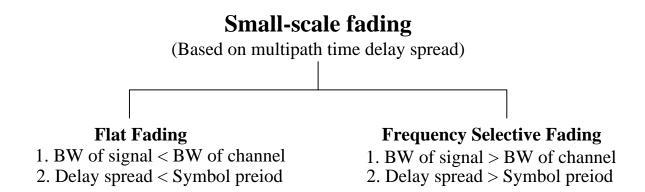
$$D(t) = \sum_{n=0}^{N-1} \left\{ a(n)\cos(2\pi f_n t) + b(n)\sin(2\pi f_n t) \right\}$$
 (1)

where 
$$f_n = f_0 + n\Delta f$$
 and  $\Delta f = \frac{1}{N\Delta t}$ 

- \* Using a two-dimensional digital modulation format, the data symbols d(n) can be represented as a(n) + jb(n)
  - \* a(n): in-phase component
  - \* b(n): quadrature component



- \* The serial data elements spaced by  $\Delta t$  are grouped and used to modulate N carriers. Thus they are frequency division multiplexed.
- \* The signaling interval is then increased to  $N\Delta t$ , which makes the system less susceptible to channel delay spread impairments.







Consider a set of transmitted carriers as follows:

$$\psi_n(t) = e^{j2\pi \left(f_0 + \frac{n}{N\Delta t}\right)t}$$
 for  $n = 0, 1, ..., N - 1$  (2)

$$\int_{a}^{b} \psi_{p}(t) \psi_{q}^{*}(t) dt = \begin{cases} (b-a) & \text{for } p = q \\ 0 & \text{for } p \neq q \text{ and } (b-a) = N\Delta t \end{cases}$$



$$\int_{a}^{b} \psi_{p}(t) \psi_{q}^{*}(t) dt = \int_{a}^{b} e^{j2\pi(p-q)\frac{t}{N\Delta t}} dt$$

$$= \frac{e^{j2\pi(p-q)\frac{b}{N\Delta t}} - e^{j2\pi(p-q)\frac{a}{N\Delta t}}}{j2\pi(p-q)/N\Delta t}$$

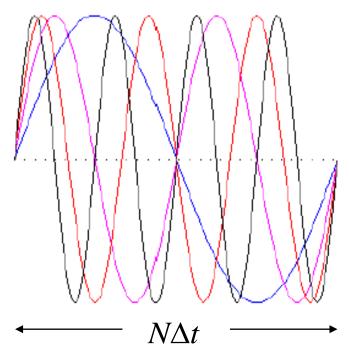
$$= \frac{e^{j2\pi(p-q)\frac{b}{N\Delta t}} \left(1 - e^{j2\pi(p-q)\frac{1}{N\Delta t}(a-b)}\right)}{j2\pi(p-q)/N\Delta t}$$

$$= 0, \text{ for } p \neq q \text{ and } (b-a) = N\Delta t$$

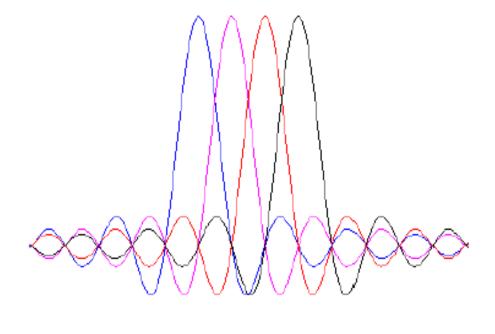


#### Time domain

Frequency domain



Example of four subcarriers within one OFDM symbol



Spectra of individual subcarriers



# Mathematical Expression of OFDM Signal

\* From above, we know that  $\{\psi_n(t)\}$  is the orthogonal signal set. An OFDM signal based on this orthogonal signal set can be written as:

$$x(t) = \operatorname{Re}\left\{\sum_{k=-\infty}^{\infty} \sum_{n=0}^{N-1} d_{k,n} \psi_n(t-kT)\right\}$$
(3)

where  $\psi_n(t) = e^{j2\pi f_n t}$  for n = 0, 1, 2, ..., N-1  $0 \le t \le T$ 

$$f_n = f_0 + \frac{n}{T} , \quad T = N\Delta t$$

$$d_{k,n} = a_{k,n} + jb_{k,n}$$



# Mathematical Expression of OFDM Signal

- \* T: OFDM symbol duration
- \*  $d_{k,n}$ : transmitted data on the *n*-th carrier of the *k*-th symbol

$$x(t) = \text{Re}\left\{\sum_{k=-\infty}^{\infty} \sum_{n=0}^{N-1} C_{k,n} \psi_n(t-kT)\right\}$$

$$= \sum_{k=-\infty}^{\infty} \sum_{n=0}^{N-1} \left\{a_{k,n} \cos(2\pi f_n(t-kT)) - b_{k,n} \sin(2\pi f_n(t-kT))\right\}$$
(4)

\* If there is only one OFDM symbol (i.e. k = 0), it can be simplified as:

$$x(t) = \sum_{n=0}^{N-1} \left\{ a_n \cos(2\pi f_n t) - b_n \sin(2\pi f_n t) \right\}$$
 (5)





According to the structure of Tx, it must use N oscillators. That increases the hardware complexity.

\* The equivalent method is using IDFT (IFFT).



In general, each carrier can be expressed as:

$$S_c(t) = A_c(t)e^{j(2\pi f_c t + \phi_c(t))}$$
(6)

\* We assume that there are N carriers in the OFDM signal. Then the total complex signal  $S_s(t)$  can be represented by:

$$S_s(t) = \frac{1}{N} \sum_{n=0}^{N-1} A_n(t) e^{j(2\pi f_n t + \phi_n(t))}$$
 (7)

where  $f_n = f_0 + n\Delta f$ 

and  $A_n(t)$ ,  $\phi_n(t)$ ,  $f_n$  are amplitude, phase, carrier frequency of n-th carrier, respectively.



\* Then we sample the signal at a sampling frequency  $1/\Delta t$ , and  $A_n(t)$  and  $\varphi_n(t)$  becomes:

$$\phi_n(t) = \phi_n \tag{8}$$

$$A_n(t) = A_n \tag{9}$$

$$S_{s}(k\Delta t) = \frac{1}{N} \sum_{n=0}^{N-1} A_{n} e^{j(2\pi(f_{0} + n\Delta f)k\Delta t + \phi_{n})}$$
(10)

Then the sampled signal can be expressed as:

$$S_{s}(k\Delta t) = \frac{1}{N} \sum_{n=0}^{N-1} \left( A_{n} e^{j(2\pi f_{0}k\Delta t + \phi_{n})} \right) \cdot e^{j2\pi nk\Delta f\Delta t}$$
(11)



\* The <u>inverse discrete Fourier transform</u> (IDFT) is defined as the following:

$$f(k\Delta t) = \frac{1}{N} \sum_{n=0}^{N-1} F(n\Delta f) e^{j2\pi nk/N}$$
 (12)

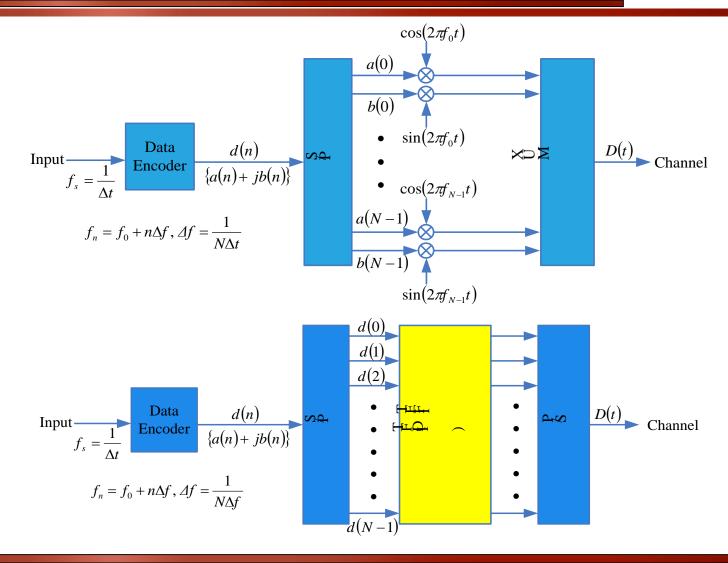
\* Comparing eq.(11) and eq.(12), the condition must be satisfied in order to make eq.(11) an inverse Fourier transform relationship:

$$\Delta f = \frac{1}{N\Delta t} \tag{13}$$



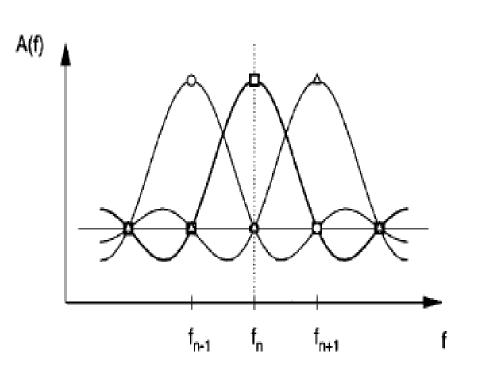
- If eq.(13) is satisfied,
  - \*  $A_n e^{j(2\pi f_0 k \Delta t + \phi_n)}$  is the frequency domain signal
  - \*  $S_s(k\Delta t)$  is the time domain signal
  - \*  $\Delta f$  is the sub-channel spacing
  - \*  $N\Delta t$  is the symbol duration in each sub-channel
- \* This outcome is the same as the result obtained in the system of Figure 1. Therefore IDFT can be used to generate an OFDM transmission signal.

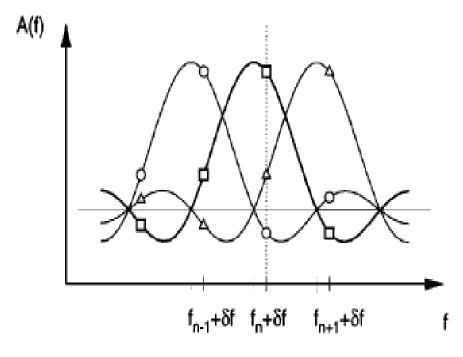






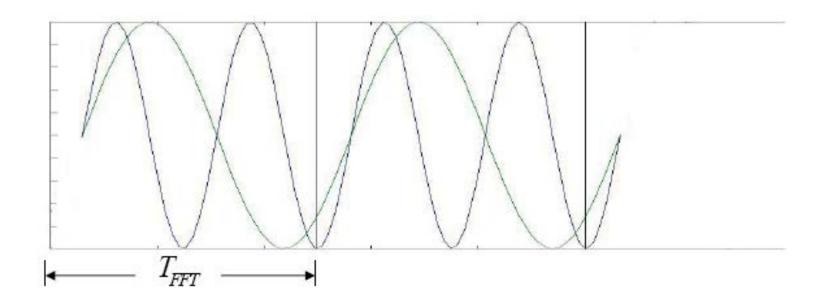
# Frequency Error Results in ICI







## Synchronization Error Results in ICI



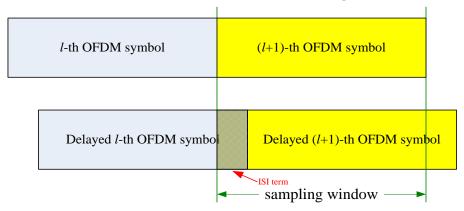
Not Orthogonal Any More.

# Cyclic Prefix (CP)



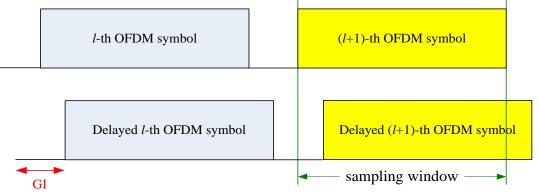


In multipath channel, delayed replicas of previous OFDM signal lead to ISI between successive OFDM signals.



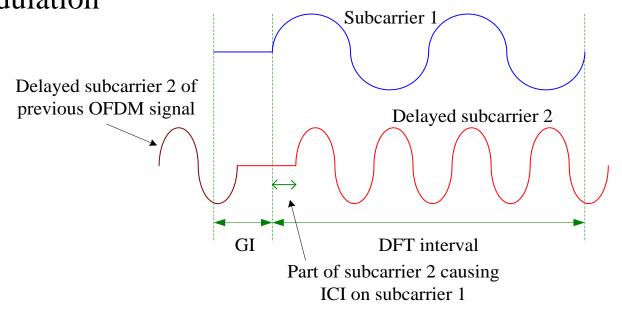
Solution: Insert a guard interval between successive OFDM

signals.





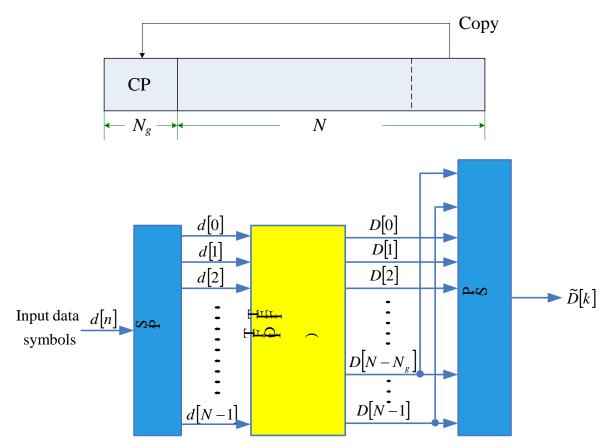
Guard interval leads to intercarrier interference (ICI) in OFDM demodulation



- In DFT interval, difference between two subcarriers does not maintain integer number of cycles → loss of orthogonality.
- Delayed version of subcarrier 2 causes ICI in the process of demodulating subcarrier 1.

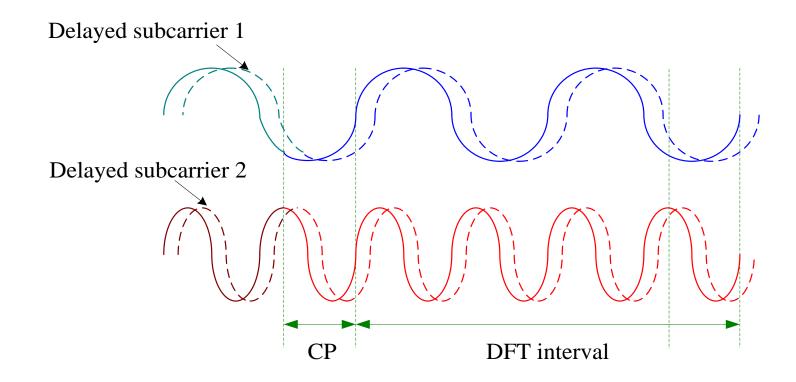


Cyclic prefix (CP): A copy of the last part of OFDM signal is attached to the front of itself.



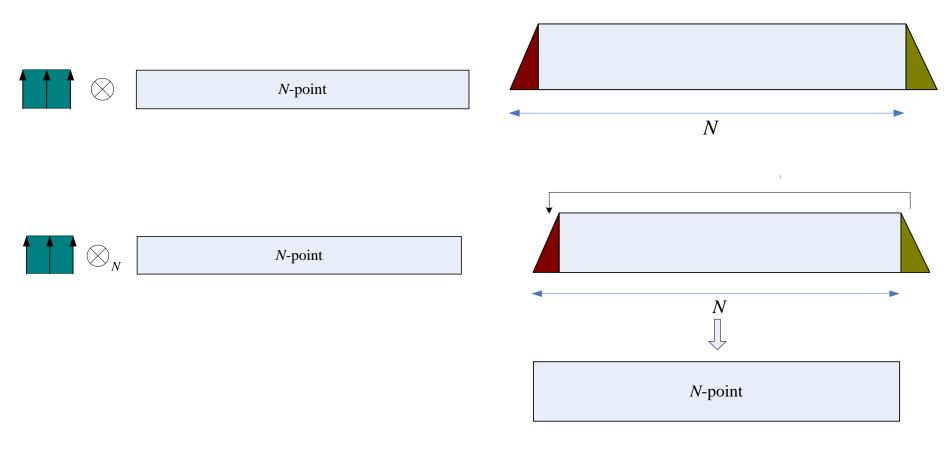


★ All delayed replicas of subcarriers always have an integer number of cycles within DFT interval → no ICI



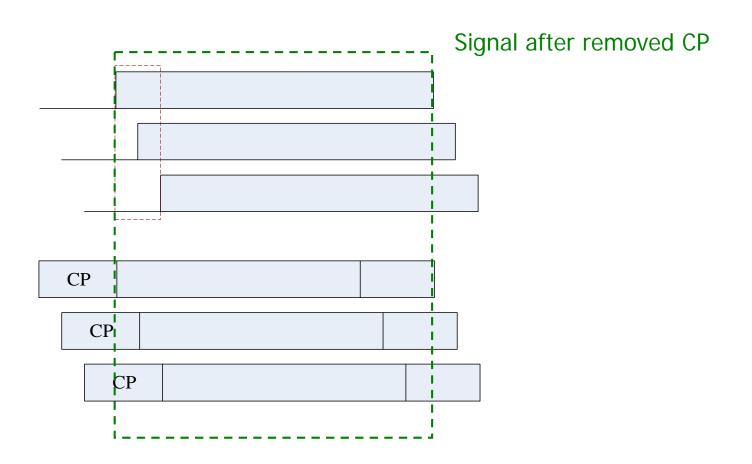


\* Linear convolution vs. circular convolution



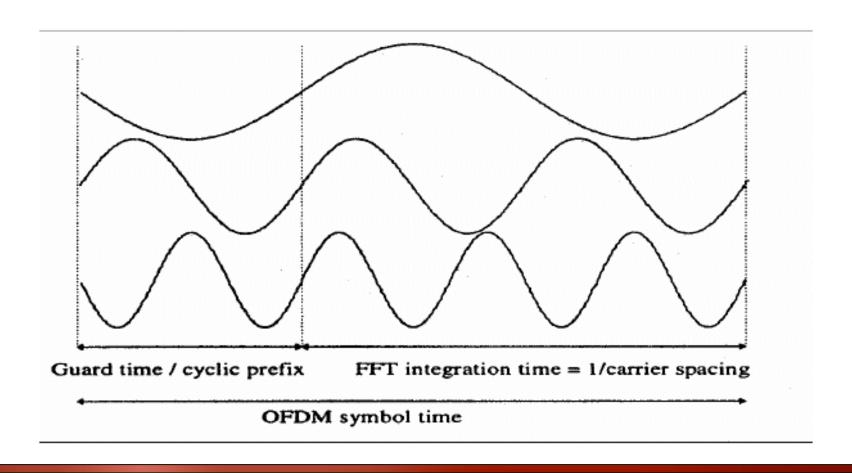


Channel effect with cyclic prefix

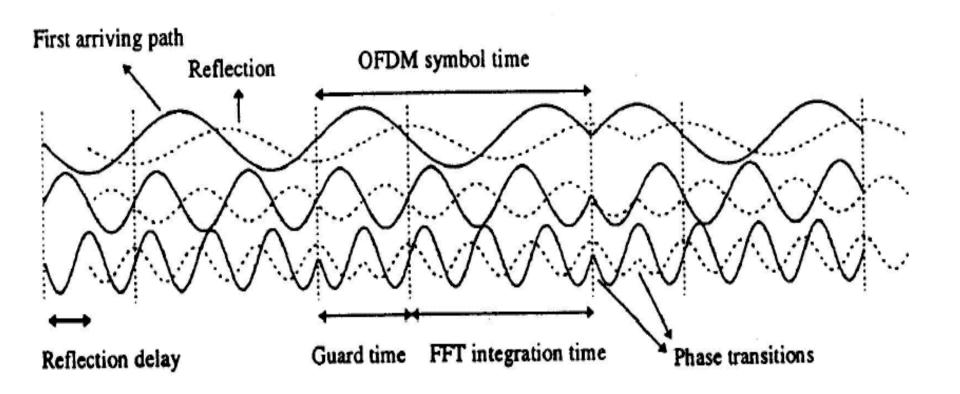




#### Time-Domain Explanation









\* Spectrum of channel response h[n] with length  $L_h$  (smaller than  $N_{\rho}$ )

$$H_k = FFT\{h[n]\}$$

Received complete OFDM signal

$$\widetilde{r}[n] = \widetilde{D}[n] \otimes h[n], \ 0 \le n \le N + N_g + L_h - 2$$

\* Received useful part r[n]

$$r[n] = D[n] \otimes_N h[n]$$

where  $\bigotimes_N$  is N-point circular convolution (due to CP)

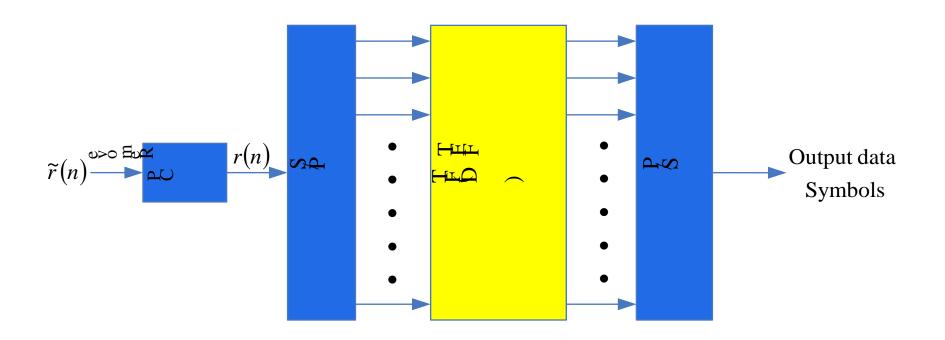
Received symbol at k-th subcarrier

$$Y_k = FFT\{r[n]\} = FFT\{D[n] \otimes_N h[n]\} = X_k H_k$$

$$\Rightarrow X_k = \frac{Y_k}{H_k}$$
 "Useful property for OFDM system to reduce complexity of channel equalization"



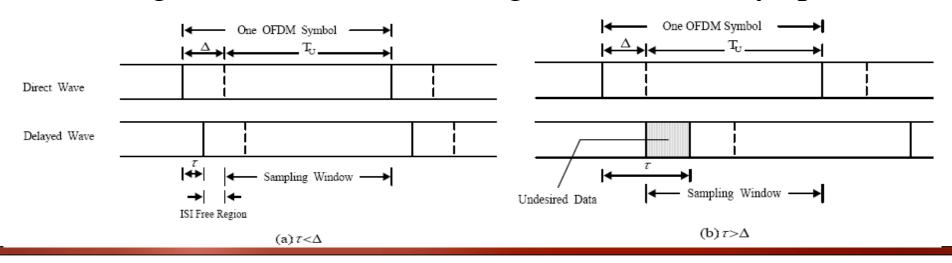
# Cyclic Prefix (OFDM Receiver)





- One of the most important reasons to do OFDM is the efficient way it deals with multipath delay spread.
- \* To eliminate inter-symbol interference (ISI) almost completely, a guard time is introduced for each OFDM symbol.

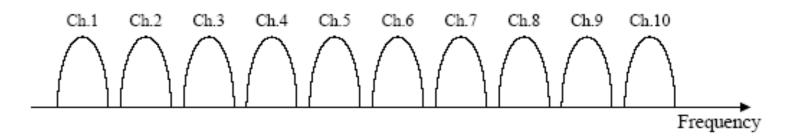
(The guard time is chosen larger than the delay spread)



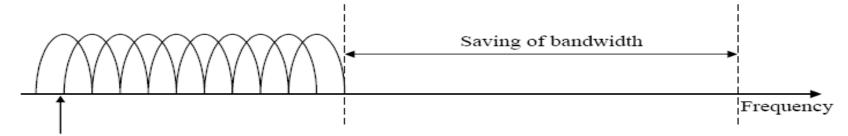


## Bandwidth Efficiency

\* In a classical parallel system, the channel is divided into *N* non-overlapping sub-channels to avoid <u>inter-carrier</u> interference (ICI).



The diagram for bandwidth efficiency of OFDM system is shown below:





#### Summary

- The advantage of the FFT-based OFDM system :
  - The use of IFFT/FFT can reduce the computation complexity.
  - \* The orthogonality between the adjacent sub-carriers will make the use of transmission bandwidth more efficient.
  - The guard interval is used to resist the inter-symbol interference (ISI).
  - \* The main advantage of the OFDM transmission technique is its high performance even in frequency selective channels.
- The drawbacks of the OFDM system :
  - It is highly vulnerable to synchronization errors.
  - Peak to Average Power Ratio (PAPR) problems.